

**Problem 1:** In this problem we again consider the problem of bandwidth scaling in a multi-path fading channel. As in lecture, consider dividing a multi-path fading channel into a set of  $L$  parallel channels, each in a different coherence band. Then we consider a discrete-time model and assume that each time sample corresponds to a different coherence-time. For a given coherence-time, let  $\mathbf{Y} = (y_1, \dots, y_L)^T$  be the complex vector of received signals, where  $y_l$  represent the received signal in the  $l$ th coherence band. Likewise let  $\mathbf{X}$  be  $L$  dimensional vector representing the channel input in each coherence band. These are related by

$$\mathbf{Y} = \mathbf{H}\mathbf{X} + \mathbf{W},$$

where  $\mathbf{H}$  is a  $L \times L$  diagonal matrix with the  $l$ th diagonal entry  $h_l$  corresponding to the channel gain in the  $l$ th coherence band, and  $\mathbf{W} = (w_1, \dots, w_L)^T$  represents the additive noise. Assume that  $\{h_l\}$  are i.i.d. complex Gaussians, with distribution  $CN(0, 1)$  and  $\{w_l\}_l$  are also i.i.d. complex Gaussians with distribution  $CN(0, N_o)$ . Finally, assume that the input has a power constraint given by

$$\sum_{l=1}^L \mathbb{E}(|x_l|^2) \leq P.$$

The capacity of this channel (assuming no CSI at the receiver or transmitter) is given by the maximum of  $I(\mathbf{X}; \mathbf{Y})$ , over all probability distributions on the input  $\mathbf{X}$  that satisfy the power constraint.

- a. Using that  $I(\mathbf{H}, \mathbf{X}; \mathbf{Y}) = I(\mathbf{H}; \mathbf{Y}|\mathbf{X}) + I(\mathbf{X}; \mathbf{Y})$ , show that

$$I(\mathbf{X}; \mathbf{Y}) \leq L \log \left( 1 + \frac{P}{LN_o} \right) - \sum_{l=1}^L \mathbb{E} \log \left( 1 + \frac{|x_l|^2}{N_0} \right).$$

- b. Assume the transmitter must allocate equal power to each coherence band, i.e.,  $\mathbb{E}(|x_l|^2) = P/L$  for all  $l = 1, \dots, L$ . Using the bound in part (a), show that the capacity of the channel (with this power allocation) must go to zero as  $L \rightarrow \infty$ .
- c. Now assume that the transmitter uses “peaky” signalling and transmits with power  $\lambda$  in only one of the frequency bands. In this case show that:

$$\lim_{L \rightarrow \infty} I(\mathbf{X}; \mathbf{Y}) \geq \frac{\lambda}{N_o} - \log \left( 1 + \frac{\lambda}{N_o} \right).$$

- d. As in lecture, now assume that the transmitter only transmits a fraction of the time, given by  $\theta$ . Show that as  $\theta \rightarrow 0$ , the mutual information converges to  $\frac{P}{N_0}$  nats/channel use.

**Problem 2:** This problem reviews some facts about complex random vectors. Let  $\mathbf{X}$  be a  $n$ -dimensional complex random vector, i.e., a vector where each entry is a complex random variable, with a given joint distribution.  $\mathbf{X}$  is *isotropically distributed* if  $U\mathbf{X}$  has the same probability density as  $\mathbf{X}$  for any (deterministic) unitary matrix  $U \in \mathbb{C}^{n \times n}$  (Recall a square matrix is unitary if  $U^*U = I$ , where  $U^*$  denotes the conjugate transpose).  $\mathbf{X}$  is Gaussian if its components are jointly Gaussian random variables.  $\mathbf{X}$  is *proper* if its pseudo covariance matrix  $E\mathbf{X}\mathbf{X}^T$  is the all zero matrix. The distribution of a proper, complex Gaussian random vector with covariance matrix  $Q$  and mean  $\mathbf{m}$  is denoted  $CN(\mathbf{m}, Q)$ . If  $\mathbf{X}$  has distribution  $CN(\mathbf{m}, Q)$  with  $Q$  non-singular, then its pdf is given by

$$f_{\mathbf{X}}(X) = \det(\pi Q)^{-1} \exp(-(X - \mathbf{m})^* Q^{-1} (X - \mathbf{m})).$$

- a. Suppose that each entry of  $\mathbf{X}$  is i.i.d. with distribution  $CN(0, \sigma^2)$ . Show that  $\mathbf{X}$  is isotropically distributed.
- b. Suppose that  $\mathbf{X}$  has distribution  $CN(\mathbf{0}, Q)$  and is isotropically distributed. Show that the entries of  $\mathbf{X}$  are i.i.d. with distribution  $CN(0, \sigma^2)$ . (recall a covariance matrix is always nonnegative definite and Hermitian.)